

Computer Aided Design and Fabrication of Energy Efficient Single Phase Induction Motor



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Submitted by

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Electronics Engineering of Bharathiar University, Coimbatore

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Certificate

This is to certify that the Project Report entitled
**Computer Aided Design and Fabrication of
Energy Efficient Single Phase
Induction Motor**

has been submitted by

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In partial fulfilment for the award of the Degree of
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during the academic year 1995-96

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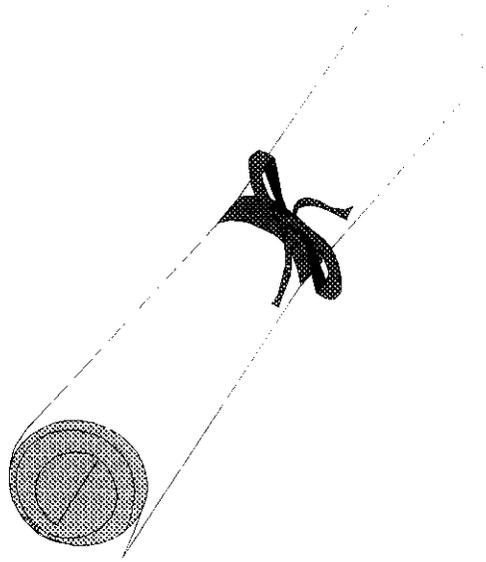
Certified that the candidate with University Reg. No. _____
was examined in Project Work Viva Voce by us on _____

Internal Examiner

External Examiner

*Dedicated to Our
Beloved Parents*





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CERTIFICATE
● — — — — —
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Date: 19.03.96

CERTIFICATE

This is to certify that the following final year students of Kumaraguru college of Technology successfully carried out their project in our industry, titled "COMPUTER AIDED DESIGN & FABRICATION OF ENERGY EFFICIENT SINGLE PHASE INDUCTION MOTOR".

<u>NAME</u>	<u>ROLL NO</u>
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2. R. Rameswaran	92EEE23
3. K.P.Thiyagarajan	92EEE33
4. R.S.Ramkumar	91EEE20

They had shown keen interest in designing the motor.

We wish them all success in their future career.



C.MUTHUSAMI B.E.,AIE.,
Managing Partner.



STARK MOTORS

Stark

7-D, SITRA ROAD, KALAPPATTI, COIMBATORE-35.

ROUTINE TEST REPORT - IS : 996

PLATE DATA

TYPE	KW	RPM	VOLT	CYS	FLC	INS. CL	RATING	PHASE
TEFC CS&R	0.185	1440	240	50	1.4	E	S ₁	SINGLE PHASE

16	Winding Resistance	Electrical Specn.
	incold at <u>34</u> °C <u>19.2</u> ohms	<u>153996</u>
	at 75°C <u>22.13</u> ohms	

Operation : NCT APPLICABLE

LOAD TEST at 240 V 50 Hz

	Current	Watts	Speed
NO LOAD	0.5	40	1470

away Starting Torque

$$W_1 \times \frac{\text{Arm Length}}{\text{Length}} = (2 \times 0.1) \times \frac{(240)^2}{(220)} = 0.24 \text{ kgm. } 200\%$$

$$\text{ut Torque} = \frac{(240)^2}{(240)} \times 9 = 9 \text{ kg. } 300\%$$

$$\text{b Torque} = \frac{(240)^2}{(240)} \times 5 = 5 \text{ kg. } 200\%$$

tested by :

R. Ramesh Kumar
R. Ramesh Kumar
S. Ramesh Kumar

Checked by :

S. Ramesh Kumar

Approved by :

STARK MOTORS

7-D, Sitra Road, Kalappatti, Coimbatore - 641 035

TEST REPORT

PERFORMANCE CALCULATION

No Load Input (W_0)	=	<u>40</u>	Watts
No Load CU. Loss (CU_0) =	$I_0^2 \times R_{75}$	=	<u>—</u> Watts
Fixed Losses	=	$W_0 - CU_0$	= <u>—</u> Watts
FL. CU. Loss	=	$I_1^2 \times R_{75}$	= <u>—</u> Watts
Total Losses	=	(3) + (4)	= <u>—</u> Watts
FL. Input	=	<u>230</u>	Watts
Stator Output	=	(6) - (5)	= <u>—</u> Watts
Rotor Output	=	(7) $\times \frac{NA}{NS} \times 0.995$	= <u>185</u> Watts
Efficiency	=	$\frac{(8)}{(6)} \times 100$	= <u>80.4</u> %
Power Factor	=	$\frac{(6)}{\cancel{230} \times I_L}$	= <u>0.7692</u>
Efficiency X P.F.	=	(9) X (10)	= <u>0.618</u>
Slip	=	$\frac{NS - NA}{NS} \times 100$	= <u>—</u> %

D. Ramakrishnan
Incharge to
S. Ramakrishnan



Acknowledgement

ACKNOWLEDGEMENT

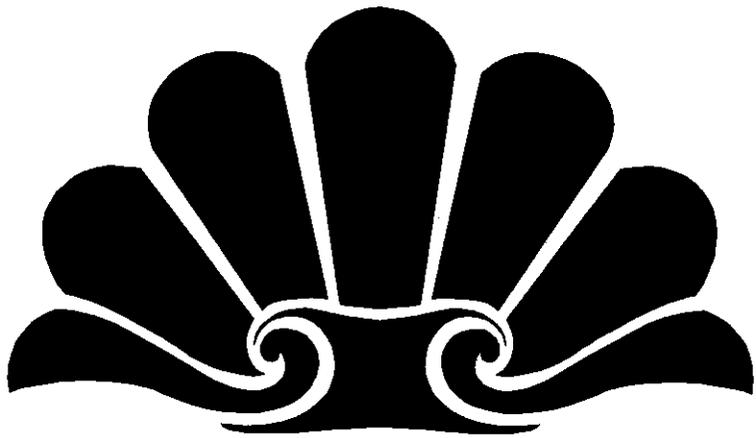
We express our sincere thanks to our Project Guide, **Mrs.N.KALAIARASI B.E, M.I.S.T.E.**, for her able guidance.

Our sincere thanks are due to our beloved Professor and Head Of The Department **Dr.K.A.PALANISWAMY B.E,M.Sc(ENGG),Ph.D., M.I.S.T.E.,C.ENG(I),F.I.E.**, for his valuable suggestions and immense help in making our project work a success.

We express our thanks to our Principal **Dr.S.SUBRAMANIAM, B.E., M.Sc(ENGG), Ph.D., SMIEEE.**, for providing all facilities in the college.

We whole heartedly express our gratitude and sincere thanks to **Mr.MUTHUSWAMY, B.E.**, Managing Partner,**STARK MOTORS** for sponsoring our project.

We also thank our friends for sparing their time and computer to carry out our Project.



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Synopsis
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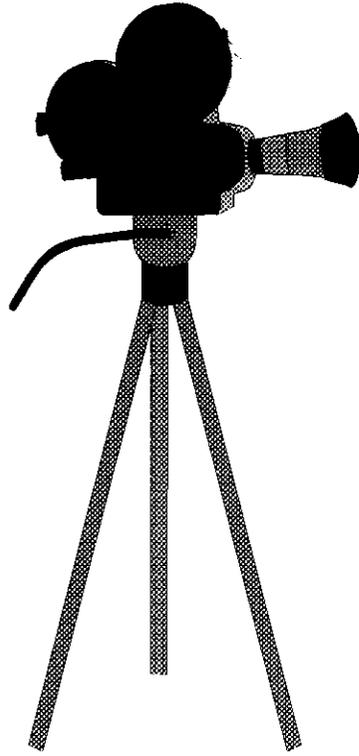
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INTRODUCTION

C H A P T E R - 1

INTRODUCTION

These motors are very widely used in industries especially in the fractional kilowatt range. They are extensively used for electric drives for low power constant speed apparatus such as machine tools, domestic apparatus and agricultural machinery. There is a large demand for single phase induction motors in sizes ranging from a fraction of a kilowatt power upto about 3.7 K.W.

Constructionally this motor is more or less similar to a polyphase induction motor except that

- (1) Its stator is provided with a single phase winding.
- (2) A centrifugal switch is used in some types of motors in order to cut out a winding used for starting purpose only.

It has a distributed stator winding under squirrel cage rotor.

1.1 TYPES OF SINGLE PHASE MOTORS

Single phase induction motors may be classified as follows depending on their construction and method of starting:

- (1) Induction motors(Split Phase, Capacitor and Shaded Pole etc.).
- (2) Repulsion motors (also called inductive-series motors).
- (3) A.C.Series Motors.
- (4) Unexcited Synchronous Motors.

1.2 ✓ **ADVANTAGES OF CAD**

Literally it is a known fact that electrical technology has developed amazingly in the recent past. But recently it has been chipped and shaped attractively to the final stage. The incoming of computers in the field of design is yet another land mark in the field of electrical technology.

The concept of CAD, the modern trend in the design field, has emerged as a boon to the design engineers. It enables creations followed by rigorous testing of graphic models without involving the chores of making physical models.

The design procedure resides as a very general software which is user alterable. The user can alter the design as and when he wanted according to any given specifications. It is also possible for him to emphasis on any part of design according to the needs.

SYNOPSIS

Energy Conservation is the foremost talk of the day and efforts are being made to bring down Power Consumption. In this project an attempt is made to design and fabricate an Energy Efficient Single Phase, 0.25 H.P, 1440 rpm Induction Motor with the help of CAD.

Program is written in C language for the design purposes.

To increase the efficiency the Conductor Area and the Core Area are increased and higher grade of Steel is used. Also higher grade insulation is used to reduce the weight of the motor.

The motor is tested and the test reports are presented in the report.

This vital advantage of flexibility in the design procedure is evident when modifications and alterations are sought after. Apart from the flexibility, the incredible speed with which the complex mathematical manipulations are done adds to the versatility.

1.3 ✓ **ADVANTAGES OF INDUCTION MOTORS**

The Single Phase Induction Motors are useful for small outputs . It has the following main advantages.

- (1) Its cost is low and very reliable.
- (2) It is very simple and extremely rugged, almost unbreakable in construction.
- (3) It requires a minimum of maintenance.
- (4) In normal operating conditions no brushes are needed hence frictional losses are reduced.
- (5) It starts up from rest and needs no extra starting motor and has not to be synchronised.

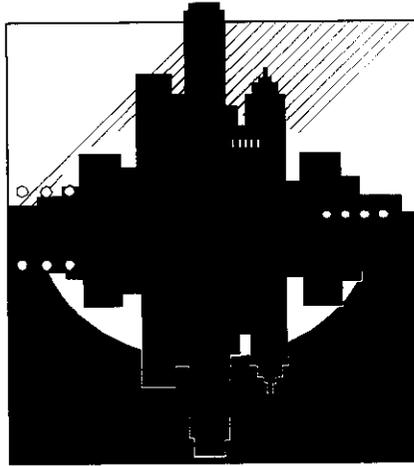
1.4 ✓ **APPLICATIONS OF SINGLE PHASE INDUCTION MOTOR**

(A) DOMESTIC APPLICATIONS:

- (1) Wet Grinders.
- (2) Blowers.
- (3) Fans.
- (4) Small pumps (Ex: jet pump).
- (5) Conveyers.

(B) INDUSTRIAL APPLICATIONS:

Petro-chemical, Cement, Rubber, Textile, Food processing, Sugar-cane Industries use single phase induction motors. In these kind of industries such motors are used for pumping purposes.



— — — — —
CONSTRUCTIONAL

DETAILS
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C H A P T E R - 2

CONSTRUCTIONAL DETAILS

2.1 INTRODUCTION

The construction of Single phase induction motor is similar to that of a three phase induction motor. However they consist essentially of a stator with running and starting windings, a squirrel cage rotor, an automatic switch to disconnect the starting winding when the motor has picked up speed and a body and end shield to which these parts, and also the terminal box, bearings etc. are fitted.

2.2 STATOR

The stator is the stationary part of the induction motor. It is a cylindrical structure, made up of laminations mounted in a cast iron or die cast aluminium frame. The stator has tapered slots with parallel sided teeth. The slots house the starting and running windings. The centre circles are used for punching rotor laminations. The laminations are either 0.35mm or 0.5mm thick.

The stator laminations are welded at several places around the outer cylindrical surface and the stack is later pushed into a frame for assembly. The segments are held together by axial key bars fitting into dovetailed slots in the outer rim of the core. The total number of segments is chosen in such a way so as to provide an equal number of joints in the core flux paths of alternating poles.

Long cores are divided into a number of stacks with radial ventilating ducts in between in order to facilitate the cooling. The width of a single stack of core should be in between 50mm and 60mm.

2.3 STATOR FRAME

Frames of electrical machines are structures in which stator core is assembled. They serve four distinct purposes:

- (1) They enclose the core and windings.
- (2) They shield the live and moving machine parts from human contact and from injury caused by intruding objects or weather exposure.
- (3) They transmit the torque to the machine supports, and are therefore designed to withstand twisting forces and shocks.
- (4) They serve as ventilating housing or means of guiding the coolant into effective channels.

A great variety of designs is employed to meet the above requirements, and to adapt machines to particular service conditions.

Frames of small machines are made as a single unit. They are provided with feet by which they are fixed to the base plate. The frames of totally enclosed machines are provided with axial fins in order to increase the heat dissipating surface.

For induction motors the frame should be strong and rigid both during construction and after assembly of the machine. This is because the length of air gap is small and if the frame is not rigid, the rotor will not remain concentric with stator giving rise to unbalanced magnetic pull.

2.4 STATOR WINDINGS ✓

Single phase induction motors are generally wound with concentric windings. They have two distinct windings. They are:

- (1) Main winding (or) Running winding.
- (2) Auxillary windings (or) Starting windings.

Each is a single layer winding, with the main winding being at the bottom of the slots. In three phase motors equal slot space is devoted to all phases but in single phase motors considerably more slot space given to main winding and less to auxillary winding. The most popular kinds of single phase windings are the concentric, progressive and skein.

Concentric windings are so named because all the coils for a single pole have a common centre and different pitches for each individual coil. That is the individual coils in the pole group are concentric coils of varying sizes.

2.5 ROTOR

The rotor consists of a block of slotted laminations mounted on a shaft. The slots form a series of tunnels when the rotor is assembled. The tunnels are filled with aluminium poured in the molten state. The bars, end rings and fan blades form one homogeneous casting. In some motors copper bars and copper end rings are used, the former being brazed into the latter. The rotor slots are skewed so that a quieter operation is obtained. In large machines the radial ventilating ducts provided in the rotor are equal to that of the stator.

2.6 SHAFTS AND BEARINGS ✓

The air gap of an induction motor is made as small as possible. Therefore the shaft is made short and stiff in order that the rotor may not have any significant deflection, as even a small deflection would create large irregularities in the air gap which would lead to production of an unbalanced magnetic pull.

Ball and Roller Bearings are generally used for low capacity motors. In single phase motors roller bearings at the driving end and a ball bearing in the non driving end are used.

2.7 STARTING SWITCHES ✓

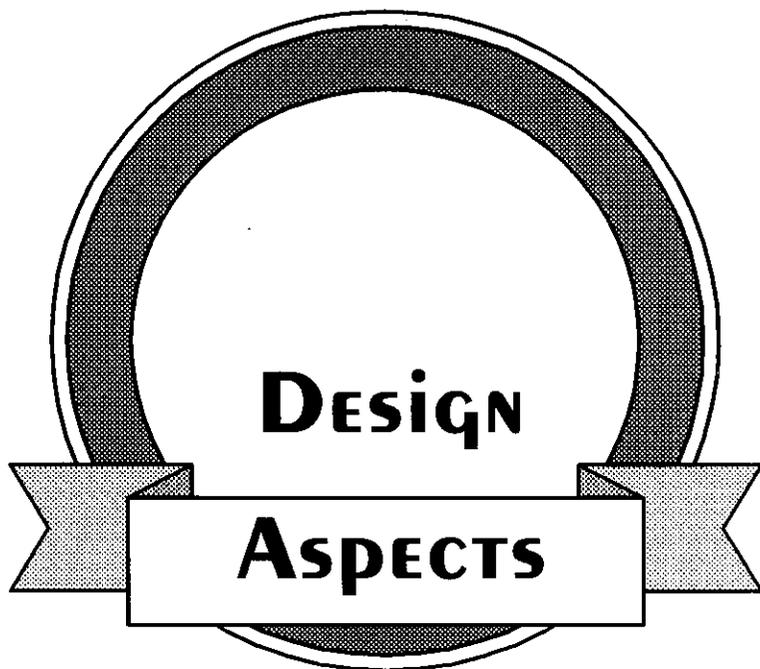
These switches are required to cut the auxillary winding out of circuit when the

motor has attained about 75% of synchronous speed. Starting switches are centrifugally operated whenever possible. For most applications, the centrifugal switch is preferable over the magnetic switch because it is not affected by the line voltage, is more reliable and is less expensive.

2.8 **ELECTROLYTIC CAPACITOR** ✓

Modern capacitor start motors employ A.C ELECTROLYTIC CAPACITORS. An A.C electrolytic capacitor is formed by winding two sheets of etched aluminium foil, separated by two layers of insulating paper into a cylindrical shape. The unit is impregnated with an electrolyte usually ethylene glycolol or a derivative. An anodic film is later produced on each foil by electro-chemical means.

The voltage rating of a capacitor is not necessarily same as that of the motor, it may be less, the same, or more depending upon how it is used. Power Factor of the electrolytic capacitor is generally of the order of 6 to 8%.



DESIGN

ASPECTS



CHAPTER - 3

DESIGN ASPECTS OF STATOR

3.1 OUTPUT EQUATIONS:

The Output Equation for Single Phase Machine is

$$\text{KVA Input } Q = C_0 \times D^2 \times L$$

$$\text{Output Coefficient} = C_0 = 11 \times K_{WM} \times B_{AV} \times A_C \times 10^{-3}$$

$$\text{KVA Input} = \text{KW} / (\cos \Phi \times \tau)$$

$$D \times D \times L = Q / C_0$$

But normally the rating of the induction motor is given by the HORSE POWER.

Then the KVA Input is

$$Q = \text{HP} \times 0.735 / (\tau \times \cos \Phi)$$

The Horse Power, Efficiency, Power factor and Speed of the machine are specified. Therefore in order to calculate the value of $D \times D \times L$, we must evaluate Output Coefficient. But the Output Coefficient depends on the choice of the electric and magnetic loadings (i.e) Values of a_c and B_{av} .

3.2 CHOICE OF AVERAGE FLUX DENSITY IN THE AIR GAP :

The choice of average flux density depends upon the following factors.

(i) **Power Factor** :The value of flux density in the air gap should be small as otherwise the machine will draw a large magnetising current giving a poor power factor.

(ii) **Iron Loss** : An increased value of B_{av} results in increased iron loss and decreased efficiency.

(iii) **OverLoad Capacity** : An increased B_{av} value means that flux per is large and if the number of turns is less, the leakage reactance become small. So the diameter of the conductor is large which means that the maximum output or in otherwords the machine has a large OverLoad Capacity.

For 50 Hz single phase machines,the value of B_{av} normally lies between 0.35 to 0.6 Wb/m}.

3.3 MAIN DIMENSIONS :

The ratio of core length to pole pitch for various design feature is :

(i) Minimum Cost - 1.5 to 2.0

(ii) Good Power Factor - 1.0 to 1.25

(iii) Good Efficiency - 1.5

(iv) Good Overall Design - 1.0

3.4 STATOR WINDINGS :

Single layer concentric windings are generally used for stator. In the concentric windings all coils have a common centre and different pole pitch.

MAIN WINDING TURNS :

$$\text{Flux per pole mm} = B_{av} \times \pi \times D \times L / P$$

$$\text{Total Turns } T_m = E / (4.44 \times f \times \Phi_m \times k_{wm})$$

3.5 CHOICE OF AMPERE CONDUCTOR PER METRE :

(i) **Copper Loss** : A large value of ac means that a greater value of copper employed in the machine results in higher copper loss and large temperature rise.

(ii) **VOLTAGE** : A small value of ac should taken for high voltage machines as in this case the space for insulation is large.

(iii) **OVERLOAD CAPACITY**: A large value of ac would result in large number of turns. This would mean the leakage reactance of the machine would become high and will result in reduced value of overload capacity.

For single phase induction motors the value of ampere conductors normally lies between 5000 to 12000 (ac/m).

3.6 NUMBER OF STATOR SLOTS:

The following factors are considered in selecting the number of stator slots.

- (i) **Tooth Pulsation Loss:** Tooth pulsation losses and noise can be minimised by using large number of narrow slots.
- (ii) **Leakage Reactance :** With a large number of slots, the machine has higher overload capacity.
- (iii) **Magnetising Current and Iron Loss :** The use of large number of slots may result in excessive flux density in teeth giving rise to higher magnetising current and higher iron loss. The stator slot pitch,

$$\begin{aligned} Y_{ss} &= \text{gap surface / total number of stator slots} \\ &= \pi \times D / S_s \end{aligned}$$

where S_s = number of stator slots

Total number of stator conductors = $2 \times T_m$

Conductors per stator slot = $Z_{ss} = 2 \times T_m / S_s$

The number of slots per pole lies between 9 to 12.

3.7 SIZE OF STATOR SLOTS :

All the stator slots do not have the same number of conductors and some contain both the running and starting winding conductors. Starting winding conductors have small cross sectional area, and its effect upon the size is small.

Generally the running winding coil with a large number of turns will determine the size of the slots. The slot liner is usually 0.3 to 0.4(MM) thick. The ratio of insulated conductor area to the slot area should never exceed 0.5. Actually this ratio should not exceed 0.35 if the winding process is made easy.

$$\text{Area required for insulated conductors} = Z_s \times (\pi / 4) \times d_1 \times d_1$$

where Z_s is the total number of the conductors.

d_1 is the diameter of the insulated conductor in mm.

The slot area provided in the stamping is calculated by multiplying the mean width by the depth of the slot.

The average slot width,

$$W_s(\text{av}) = \pi \times (D + d_{ss}) / S_s - W_{ts}$$

where d_{ss} = depth of the stator slots

w_{ts} = width of the stator tooth

S_s = number of the stator slots.

Area of each slot = $W_{s(av)} \times d_{ss}$

3.8 STATOR TEETH :

We should check the flux density in the stator teeth in order to see that it is not excessive. The stator tooth density B_{ts} can generally be from 1.4 to 1.7 Wb/m². If the low losses and noise are important or if the motor is totally enclosed, the lower densities should be used. For general purpose machines a flux density of 1.45 Wb/m² is taken while for high torque it may go up to 1.8 Wb/m².

As stacking factor of 0.95 is taken,

Nett iron length L_i = $0.95 \times L$

Flux density in the stator teeth,

$$B_{ts} = \frac{\Phi}{(S_s/p \times L_i \times W_{ts})}$$

3.9 STATOR CORE :

The flux density in the stator core should not exceed 1.5 Wb/m².

Generally it lies between 0.9 to 1.4 Wb/m².

$$\text{Flux in the stator core} = \Phi_c = \Phi_m/2$$

Flux density in the stator core,

$$B_{cs} = \Phi_m / (2 \times L_i \times d_{cs})$$

where d_{cs} is the depth of the stator core.

$$\text{Area of the stator core} = L_i \times d_{cs}$$

Outer diameter of stator laminations,

$$D_o = D + 2(\text{depth of the stator slots} + \text{depth of core})$$

$$D_o = D + 2d_{ss} + 2d_{cs}.$$

LENGTH OF MEAN TURN :

The length of the mean turn for each of the coils per pole of a concentric winding,

$$L_{mt} = 8.4 \times (D + d_{ss}) / S_s \times \text{Slot span} + 2 \times l$$

CHAPTER - 4

DESIGN OF AIR GAP

The following factors should be considered when estimating the length of the air gap.

- (i) **Power Factor** : If the power factor increases then the length of the air gap reduces.
- (ii) **OverLoad Capacity** : The length of air gap affects the value of zig-zag leakage reactance which forms a large part of total leakage reactance in the case of induction motors. If the length of air gap is large, the zig-zag leakage flux is reduced resulting in a reduced value for leakage reactance. So the overload capacity will increase.
- (iii) **Pulsation loss** : The pulsation loss is less with larger air gaps.
- (iv) **Unbalanced magnetic pull** : If the length of air gap is small even a small deflection or eccentricity of the shaft would produce a large irregularity in the length of air gap and is responsible for the production of large unbalanced magnetic pull.
- (v) **Cooling** : If the length of air gap is large the distance between the stator and the rotor is large. This would provide a better ventilation or cooling.
- (vi) **Noise** : The noise of the motor is reduced by increasing the length of the air gap.

The following empirical relation gives the value of length of air gap in millimeter.

$$(i) \quad lg = 0.007 \times \text{rotor diameter} / \sqrt{p}$$

$$(ii) \quad lg = 0.2 + 2 \sqrt{(D \times L)}$$

$$(iii) \quad lg = 0.125 + 0.35 \times D + L + 0.015 \times Va$$

$$(iv) \quad lg = 0.2 + D$$

CHAPTER - 5

DESIGN OF SQUIRREL CAGE ROTOR

5.1 NUMBER OF ROTOR SLOTS :

The selection of number of rotor slots in squirrel cage motors is very important and a considerable attention should be paid to select a suitable value. This is because with certain combination of stator and rotor slots the machine may refuse to start or may crawl at some synchronous speed. In some cases severe vibrations may be set up generating excessive noise. The effects are produced by harmonic fields. The harmonic fields are due to :

- (i) **Windings**
- (ii) **Slotting**
- (iii) **Saturation**
- (iv) **Irregularities in the air gap.**

The difference between the stator and the rotor slots should not be equal to

- (i) $0, \pm p, \pm 2p, \pm 3p, \pm 5p$ -----> To avoid synchronous cusps.
- (ii) $\pm 1, \pm 2, \pm (p \pm 1), \pm (p \pm 2)$ -----> To avoid noise, vibration.

5.2 AREA OF THE ROTOR BARS :

The cage rotor winding may be either of copper bars and end rings or of cast aluminium. Also technical advantages lie with copper but manufacture is cheaper with cast aluminium. Also with cast rotors, the joints between bars and end rings are eliminated.

A high rotor resistance is desirable from the stand point of starting torque and current but leads to high slip and poor efficiency. The total rotor copper section is generally 0.5 to 0.8 of total stator copper section.

Total cross section of rotor bars $A_r = S_r \times ab$

where,

ab = area of each bar, mm²

Ratio A_r/A_m = 0.5 to 0.8 for copper rotors

Since aluminium, generally used for this purpose, has a resistivity approximately twice that of copper, total bar area must be two times the area required for copper.

Ratio A_r/A_m = 1.0 to 1.6 for aluminium rotors.

5.3 **SHAPE AND SIZE OF ROTOR SLOTS :**

The rotor slots for squirrel cage rotor may either be closed or semi-closed types.

Closed slots are preferred for small size machines because the reluctance of air gap is not large owing to absence of slots opening. It reduced magnetising current. As surface of rotor is smooth, the operation of machine is quieter. Advantage is that leakage reactance is large therefore the current at starting can be limited, so we can use to start with direct on line starters. But the result in reduced the overload capacity. Buy semi-closed slot gives a better over load capacity.

5.4 **ROTOR BARS :**

The rectangular shaped bars and slots are generally preferred to circular bars and slots as the higher leakage reactance of the lower part of the rectangular bars, during starting, forces most of the through the top of the bar. It gives increase rotor resistance at starting, so improves the starting torque. The value of resistance depends upon current density used for rotor conductors. Current density of rotor taken as 4 to 7 A/mm².

5.5 **END RING :**

The stator field produces emfs of fundamental frequency in the bar. The magnitude of emf in the bar are assumed to be infinitely distributed, the distribution

of emf can be considered as sinusoidal in the bars over a pole pitch. This emf produced in the bars would circulate currents. The resistance of end rings is negligible when compared with that of the bars, the resistance coming in each current path is the resistance of two bars. Thus the current which the bars carry would be proportional to their instantaneous emfs which in turn depend upon the position of the bars in the magnetic field.

The maximum value of the current in the end ring is

$$\begin{aligned}
 I_e(\max) &= \text{bars per pole}/2 \times \text{current per bar} \\
 &= S_r/2 \times I_b(\max)
 \end{aligned}$$

5.6 ROTOR TEETH :

The width of the rotor slot should be such that the flux density in the rotor teeth does not exceed about 1.7 Wb/m^2 . The maximum flux density occurs at their roots as their section is minimum.

Minimum width of the rotor teeth,

$$W_{tr}(\min) = \text{mm} / (1.7 \times (S_r/p) \times L_i)$$

Minimum width of tooth actually provided

$$\begin{aligned}
 W_{tr} &= \text{Rotor slot pitch at the root} - \text{Rotor slot width.} \\
 &= \pi (D_r - 2 \times d_{sr}) / S_r - W_{sr}
 \end{aligned}$$

Where, d_{sr} = Depth of the Rotor slot.

W_{sr} = Width of stator slot.

5.7 ROTOR CORE :

The flux density in the rotor core is generally equal to the Stator core density.

$$\text{Depth of rotor core} = \frac{\Phi_m}{2 \times B_{cr} \times L_i}$$

Where, B_{cr} = Flux density in the Rotor core.

Inner Diameter of Rotor Lamination,

$$D_i = D_r - 2 \times (d_{sr} + d_{cr})$$

The flux density in rotor teeth and core can be taken slightly higher than those in the stator teeth and core.



SOFTWARE

CHAPTER - 6

SOFTWARE

6.1 ALGORITHM

STEP :01 :-

For the design of single phase induction motor the input requirements are Rated Voltage, Horse Power, Speed, Efficiency, Power Factor, Specific Magnetic Loadings and Specific Electric Loadings.

STEP :02 :-

Winding factor of the Main winding is calculated by following equation.

Pitch factor of coils = $\sin(\text{coil span}/\text{slots per pole})$

winding factor of concentric winding is,

$K_{wm} = \frac{\text{Sum of product of pitch factor of each coil to turns in each coil}}{\text{total turns per pole}}$

STEP :03 :-

The main dimensions are obtained by the following equations.

$$(1) \quad n_s = n / 60$$

$$(2) \quad c_0 = 11 \times B_{av} \times a_c \times k_{wm} \times 10^{-3}$$

$$(3) \quad Q = \frac{P_{op} \times 10^{-3}}{(\tau \times \cos \Phi)}$$

$$(4) \quad D^2 \times L = \frac{Q}{(c_0 \times n_s)}$$

Core diameter can be obtained upon the Length to Pole pitch ratio.

STEP :04 :-

The flux per pole in stator,

$$\Phi_m = B_{av} \times \pi \times D \times L/P$$

STEP :05 :-

Main winding parameters are designed by the following equations.

(i) Induced voltage, $E = 0.95 \times V$

(ii) Total series turns in main winding is :

$$T_s = \frac{E}{(4.44 \times K_{wm} \times f \times \Phi_m)}$$

- (iii) Total turns per pole, $T_p = T_s/P$
- (iv) Stator Rated current, $I_s = P / (V \cos \Phi \times \tau)$
- (v) Area of the main winding, $A_m = I_s / J_m$.

Where,

J_m = Current density in Main winding.

J_m is lies between 7 to 9 A/mm²

(vi) Diameter of the Main winding is, $d_m = \sqrt{A_m \times 4 / \pi}$

(vii) Length of Mean turn of main winding is

$L_{mt} = 8.4 \times (D + d_{ss}) / S_s \times \text{slot spanned} + 2 \times l$

STEP :06 :-

Winding factor of axillary winding.

pitch factor of aux winding for coils

= $\sin ((\text{coil span} + 1) / \text{slots per pole})$

$K_{wa} = \frac{\text{sum of product of pitch factor of each coil to turns in each coil}}{\text{total turns per pole in aux winding.}}$

STEP :07 :-

STARTING WINDING DESIGN :

Starting winding parameters are found out by the following equations.

(i) Total no of turns = $T_a = k \times T_s \times K_{wm} / K_{wa}$

(ii) Total turns per pole = $T_{pa} = T_a / P$

(iii) Area of the aux winding,

$$A_s = \text{Area of the main winding} / 2 \text{ to } 2.4$$

(iv) Dia of the aux winding = $d_a = \sqrt{(A_s \times 4 / \pi)}$

(v) Length of mean turn of aux winding is,

$$L_{mts} = 8.4 \times (D + d_{ss}) / S_r \times \text{slot spanned} + (2 \times L)$$

STEP : 08:-

STATOR SLOT DESIGN:

The stator slot parameter are obtained by the following steps

$$(i) \quad \text{stator slot pitch} = S_{sp} = \pi \times T / S_s$$

$$(ii) \quad \text{Depth of the stator core} = d_{sc} = \Phi_m / (2 \times L_i \times b_{cs})$$

where b_{cs} is the flux density in the stator core (0.92 to 1.4 wb/m²)

$$(iii) \quad \text{Width of the stator teeth} = W_{st} = \Phi_m / ((s_s/p) \times L_i \times b_{st})$$

$$\text{where } L_i = 0.95 \times L$$

$$b_{st} = \text{flux density in the stator (1.4 to 1.7 wb/m)}$$

$$(iv) \quad \text{Slot area} = \text{conductor area/ space factor}$$

(Normally the space factor = 0.5)

$$\text{Conductor area} = 2 \times T_m \times A_m$$

STEP : 09:-

$$\text{Length of air gap} = l_g = 0.2 + 2 \times \sqrt{(D \times L)} \text{ mm}$$

where D & L are in metres.

STEP : 10 :-

DESIGN OF SQUIRREL CAGE ROTOR:

Normally Squirrel cage rotors are used in the single phase induction motors.

The values are calculated by the following eqns.

(i) Rotor outer diameter D_r = stator inner dia - 2 x length of the air gap

$$D_r = D - 2 \times l_g$$

(ii) Total conductor section in stator,

$$= T_{cu} = 2 \times T_m \times A_m$$

(iii) (a) **Copper rotor:**

$$\text{Total conductor section in the rotor} = 0.55 \times T_{cu}$$

(b) **Aluminium rotor:**

$$\text{Total conductor section in the rotor} = 1.1 \times T_{cu}$$

STEP : 11 :-

ROTOR SLOT AND BAR DESIGN :

The rotor slot parameters are obtained from the following equations.

(i) Rotor Slot Pitch = $\pi \times dr / Sr$

(ii) Width of the Rotor Teeth,

$$W_{tr} = \Phi_m / ((Sr/p) \times L_i \times b_{tr})$$

where b_{tr} = flux density in the rotor teeth. Its value is slightly more than the flux density in the Stator Teeth(b_{ts}).

(iii) Depth of the Stator Core,

$$d_{sc} = \Phi_m / (2 \times L_i \times b_{cr})$$

(iv) Bar Current I_b = $2 \times i \times \cos \Phi \times ts \times k_{wm} / Sr$

(v) The Bar Width and Depth are equal to the slot dimensions(excluding neckr).

(vi) The Length of each bar = $L_b = \sqrt{(L) + S_{sp}}$

The Rotor Bars are skewed by one slot pitch in the Rotor slots.

STEP : 12 :-

END RING DESIGN :

The End Ring Parameters are obtained by the following formulae.

$$(i) \quad \text{Area of the end ring} = 0.32 \times tcsr / p$$

where $tcsr$ is total conductor section in the rotor

In this case the current density in bars is assumed to be equal to the current density in the end ring.

The depth and thickness are selected depending upon the area of the end ring (Assign the value of depth in order to find the thickness of the end ring).

$$(ii) \quad \text{End ring current } I_e = S_r \times I_b / (\pi \times p)$$

STEP : 13 :-

(i) Gap Contraction of the Stator Slots,

$$K_{gs} = Y_s / (Y_s - K_{cs} \times W_o)$$

where W_o is Slot opening of Semiclosed Stator slots

$K_{cs} =$ Carter's Gap Coefficient.

$$= \frac{1}{1 + 5 \times (l_g / W_s)}$$

where W_s is the width of the Stator slots

(ii) Gap Contraction Factor of the Rotor Slots (K_{gr}) are found out in the similar manner as that of the Stator Slots.

(iii) Gap Contraction Factor of the Machine,

$$K_g = K_{gs} \times K_{gr}$$

STEP : 14 :-

RESISTANCES AND REACTANCES OF THE WINDINGS :

(i) Resistance of the Stator Windings, Rotor Bar and End Ring Resistances at 20 degrees and 75 degrees are found out by the standard formulae of the Induction Machines.

(ii) Specific Permeance of Stator, Rotor and Zig-Zag Leakage are found out by using the standard formulae of Semi Closed type slots.

- (iii) The Slot Leakage Reactance, Zig-Zag Leakage Reactance, Over-Hang Leakage Reactance and Magnetizing Reactance are found out by standard formulae of the Semi-Closed Slots type Induction machines.

STEP : 15 :-

CAPACITOR DESIGN :

- (i) Capacitive Reactance required for maximum starting torque,

$$X_c = X_{la} + (R_a + R_m) / (Z_m + X_{lm})$$

where X_{la} = Total Leakage Reactance in terms of Auxillary Winding.

R_a = Total resistance in terms of Aux wdg.

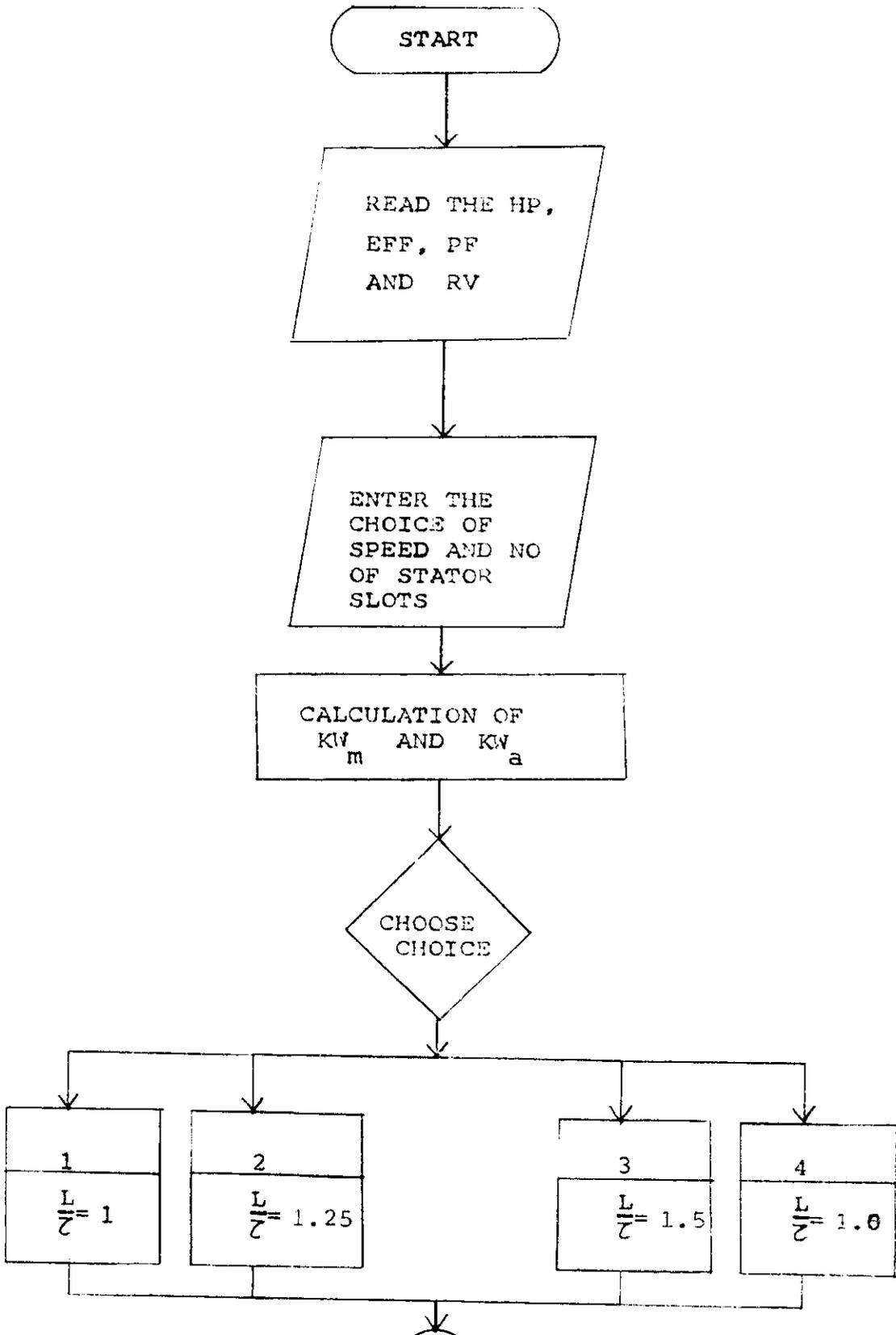
R_m = Total resistance in terms of Mainwdg

Z_m = Locked Impedance of Main winding.

X_{lm} = Total Leakage Reactance Referred to Main winding

- (ii) Capacitance $C = 10^6 / (2 \times \pi \times f \times X_c)$

- (iii) We should select the nearest and also the smaller value of the Capacitor available in the Market.



1

CALCULATION OF MAIN WINDING PARAMETERS

CALCULATION OF AUXILLARY WINDING PARAMETERS

CALCULATION OF STATOR SLOT DIMENSIONS

DESIGN OF LENGTH OF AIR GAP

CALCULATION OF MAIN DIMENSIONS

IF $Z_5 = 1$

COPPER ROTOR
ROTOR CONDUCTOR SECTION = $0.55 \times$ STATOR CONDUCTOR SECTION

ALUMINIUM ROTOR
ROTOR CONDUCTOR SECTION = $1.1 \times$ STATOR CONDUCTOR SECTION

2

DESIGN OF ROTOR
MAIN DIMENSIONS

DESIGN OF ROTOR
SLOTS AND BARS

DESIGN OF ROTOR
END RINGS

CALCULATION OF GAP
CONTRACTION FACTOR

CALCULATION OF
RESISTANCE OF THE
WINDINGS

CALCULATION OF
REACTANCE OF THE
WINDINGS

DESIGN OF CAPACITOR
AND SELECT THE
NEAREST MARKET
VALUE

3

CALCULATION OF MMF
IN THE AIR GAP

CALCULATION OF MMF
IN STATOR
AND ROTOR

CALCULATION OF
MACHINE DETAILS
UNDER LOADED
CONDITIONS

CALCULATION OF
MACHINE DETAILS
UNDER NO LOAD
CONDITIONS

TEMPERATURE RISE
CALCULATIONS

PRINT
STATOR DIMENSIONS

4

4

PRINT
STATOR SLOT DIMENSIONS

PRINT MAIN
WINDING PARAMETERS

PRINT AUXILIARY
WINDING PARAMETERS

PRINT
ROTOR DIMENSIONS

PRINT END-RING
AND BAR PARAMETERS

PRINT RESISTANCE
and REACTANCE OF
WINDINGS

PRINT CAPACITOR
VALUE AND
NEAREST MARKET
VALUE

PRINT IRON AND
COPPER LOSSES OF
MACHINE

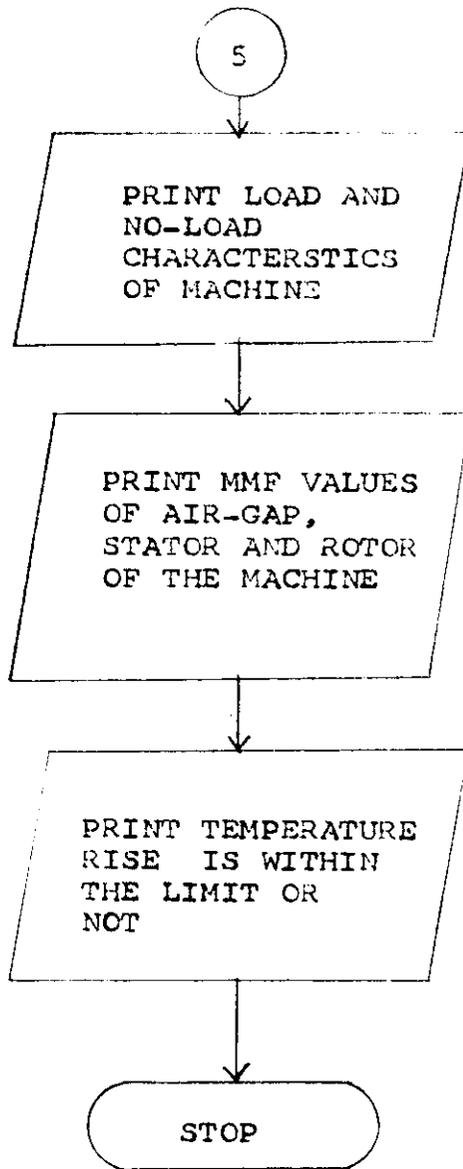
5

PRINT LOAD AND
NO-LOAD
CHARACTERSTICS
OF MACHINE

PRINT MMF VALUES
OF AIR-GAP,
STATOR AND ROTOR
OF THE MACHINE

PRINT TEMPERATURE
RISE IS WITHIN
THE LIMIT OR
NOT

STOP



```

# include <stdio.h>
# include <string.h>
# include <math.h>
# include <conio.h>
# include <graphics.h>
# include <stdlib.h>
# include <dos.h>
# define k 1.53 :
void main()
{
    int ac=10000,c3,c6[10],j,j1,j3,j4,ks1,ks2,ns,n,
    rx,rs,rv,ta,tpm,tppa,tta,ttm,x3,x4,z,z3,z5;

    float a,ab,atg,asw,aer,abl,add,aos,aa,a1,ats,atr,atcs,atcr,at60,
    atst,atrt,atsc,atrc,aeid,bav=0.4582,bcs=1.2,bcr=1.4,bst=1.45,bb,b60,
    bts1,bts60,btr60,btr1,c0,c[10],c1[10],c2[10],c4[10],c5[10],cap,c7,c8,
    ctr,cs,cus,cur,cue,cc,cx,clsp,ccd,ccis,
    d,d1,dsc,d0,dss,doe,dcr,dr,doe,dors,dero,deri,dem,dss0,
    e,eff=0.9,err,effl,ermw,fpp,fdt,fw1,fdsc,
    h,he,ist,isc,i,im,il,i0,id=7.5,ib,ie,ip,idr,ics,ldos,
    k1,k2,kr=0.96,kg,kwm,kgss,kgsr,kwa,kcs,kcr,kl,
    l,l1,lb,lg,ln1,ln2,lrcm,lrca,lmt[10],lmtm,lmtn,lmts,lip,lmta[10],ldis,
    lded,lnty,lipr,lcs,lcr,max,msc,m,ntm,nlpf,neck,neckr,
    op,opl,ocs,pf,pac,p1,p,q,q1,r,r1,rr1,rr2,rsp,rsw,rr,rs1,rs2,rrm,
    sa,ssp,spz,scr1,scr2,slht,sps,spr,sar,sct,scpf,st,spp,ss,sr,sd,tr,tlds,
    tou,tld,ttx,TTY,tsc,trst,tirs,tcsm,tcsr,tcsb,toe,tts,tim,til,trrs,trsw,tlrs,
    trmw,tl,tcm,tnl,w0,word,wtr,wsst,wors,wgs,wst,wgt,ws,wo,wrt,wts1,wtr1,
    va,vae,x,x0,xo,xc,xs,xz,xm,xcl,x6,x5,xsl,xlm,xom,y,z1,z2,za,zm;

    int    arr[6]={35,50,70,110,135,225},

           ar[5]={0,3000,1500,1000,750},

           ssa[8]={0,54,36,32,30,28,24,18},

           aw[16]={0,12,18,25,40,46,60,90,120,
                   180,250,370,550,750,1100,1500};

    float  adss[16]={0,0.003,0.004,0.005,0.007,0.008,0.009,
                    0.011,0.013,0.015,0.017,0.019,0.022,0.025,
                    0.028,0.030},

           bt[35]={0,0.3,0.4,0.5,0.6,0.7,0.8,0.9,0.95,1.0,1.05,1.1,
                  1.15,1.2,1.25,1.3,1.35,1.4,1.45,1.5,1.55,1.6,1.65,
                  1.7,1.75,1.8,1.85,1.9,1.95,2.0,2.05,2.1,2.15,2.2,2.25},

           at[35]={0,170,180,200,210,220,300,330,350,400,410,420,450,
                  500,600,700,900,1100,1450,1700,2400,2900,4000,5000,
                  7000,9000,12000,14000,19000,26000,34000,42500,
                  50000,70000,95000};

    int gd =CGA, gm=CGAC1;
    initgraph(&gd, &gm, "c:\\tc\\bgi");

```



```
c8=c8+(c2[j])*(c2[j]);
}
```

```
/** THE FOLLOWING VALUES ARE SEMI CIRCULAR SLOT VALUES***/
```

```
ws=0.145; h=0.3;
he=0.042; w0=0.03937;
```

```
cx=(c8/(c7*c7))*(1/(kwm*kwm))*(ss/(4.0*p));
sps=4*3.142857143e-7*(0.8+(neck/lip)+(2*scri1)/(lip*1e-3+(2*scri1)));
spr=(h/(3*ws)+(he/w0))*88.0/7.0*1e-7;
xs=16*3.141592654*50*ttm*ttm*kwm*kwm*cx*(sps+spr*ss/sr)*l/ss;
```

```
/** sps,spr ARE SPECIFIC PERMEANCE OF STATOR AND ROTOR SLOTS ***/
/** xs IS SLOT LEAKAGE REACTANCE REFERRED TO MAIN WINDING ***/
```

```
wsst=ssp-lip*1e-3;
wrt=rsp-lipr*1e-3;
spz=wsst*wrt*(wsst*wsst+wrt*wrt)*4*3.142857143*1e-7/(12*lg*1e-3*ssp*ssp*rsp);
xz=16*3.141592654*50*ttm*ttm*kwm*kwm*1*spz/ss;
/** spz IS SPECIFIC PERMEANCE OF ZIGZAG LEAKAGE ***/
/** xz IS ZIGZAG LEAKAGE REACTANCE ***/
```

```
x4=1;
x3=2*k2;
r1=0;
for(j=1;j<=(int)k2;j++)
```

```
{
  r1=r1+(x3-x4);
  x4++;
  x3--;
}
```

```
r=r1/(int)k2; /** r REFERS TO AVERAGE COIL SPAN***/
x0=16*3.142857143*850*ttm*ttm*kwm*kwm*(4*3.142857143e-7/(6.4*ss*p))*3.142857143*(d
```

```
/** x0 REFERS TO OVERHANG LEAKAGE REACTANCE***/
```

```
xm=16*3.142857143*50*ttm*ttm*kwm*kwm*1*4*3.142857143*1e-7*tou/(lg*1e-2*p*kg*1.15);
```

```
/** xm REFERS TO MAGNETIZING REACTANCE ***/
```

```
aos=(3.142857143/(sr/p))*1*sr/ss; /** aos REFS TO ANGLE OF SKEW**/
xsl=xm*(aos*aos/12.0)*0.95; /** xsl REFS TO SKEW LEAKAGE REACTANCE**/
xlm=xs+xz+x0+xsl; /** xlm REFS TO TOTAL LEAKAGE REACTANCE***/
xom=xm+xlm/2; /** xom REFS TO OPEN CKT REACTANCE***/
kl=sqrt((xom-xlm)/xom); /** kl REFS TO LEAKAGE FACTOR ***/
add=xs+xz+x0;
```

/*

CAPACITOR DESIGN

```
ermw=1.2*rr2; /** ermw = EFFECTIVE ROTOR RESIST IN MAIN WINDING***/
trmw=ermw+rs2; /** trmw = TOTAL RESISTANCE IN MAIN WINDING***/
trrs=(1.52*1.52)*ermw; /** trrs=TOT ROTOR RESIS IN TERMS START WINDING***/
rsw=0.017*tta*lmst/(a1*1e6);
trsw=trrs+rsw; /** trsw =TOT RESIS IN STARTING WDG***/
tlrs=(1.52*1.52)*xlm; /** tlrs = TOT LEAK REACT IN STARTING WINDING***/
zm=sqrt((trmw*trmw)+(xlm*xlm)); /** zm = IMPEDANCE OF MAIN WINDING***/
xc=tlrs+((trmw+trsw)/(zm+xlm)); /** xc REFS TO CAPACITIVE REACTANCE***/
```


6.4 LIST OF VARIABLES USED FOR SINGLE PHASE MOTOR

✓ a	=	Area of Main Winding (square metre)
✓ a1	=	Area of Secondary Winding(square metre)
aa	=	Conductor area for two windings in same slot.
ab	=	Sum of area of two semi circles in stator slots.
✓ ac	=	Ampere conductors per meter.
✓ ab1	=	Area of the bar (square metre).
add	=	Sum of the slot,zig-zag & overhang leakage reactances.
✓ aer	=	Area of the End ring(square meter).
✓ asw	=	Average slot width of stator slots(meter).
atg	=	MMF in the air gap.
ats	=	MMF in the Stator teeth.
atr	=	MMF in the Rotor teeth.
aos	=	Angle of skew in rotor bars.

aeid	=	Area of end surfaces.
ates	=	MMF in the stator core.
atcr	=	MMF in the rotor core.
at60	=	Total MMF in the Machine.
atcr	=	MMF in the stator core.
bb	=	Conductor area, only for the aux wdg in slots.
✓ bav	=	Flux density in the Air Gap.
✓ bcs	=	Flux density in the Stator core.
✓ bcr	=	Flux density in the Rotor core.
bst	=	Flux density in the Stator core.
b60	=	max flux density in air gap.
c7	=	Sum of main winding coils.
c8	=	Sum of square of the main winding coils.
cc	=	Conductor area, only for the main wdg in slots.

- ✓ c_o = Output co-efficient of the machine.
- c_x = Arbitrary correction factor
- ✓ c_{ap} = Calculated capacitor value.
- c_{cd} = Cooling co-efficient of ducts.
- c_{us} = Total copper loss in the Stator.
- c_{ur} = Total copper loss in the Rotor.
- c_{ue} = Total copper loss in the end ring.
- $c\{J\}$ = Pitch factor of each coils in Main winding.
- c_{cis} = Cooling co-efficient of Inner surface.
- c_{lsp} = Total copper Loss in stator portion.
- $c1\{J\}$ = % of turns per pole in each coils in Main wdg.
- $c2\{J\}$ = Total series turns in each coil in Main wdg.
- $c4\{J\}$ = Pitch factor of Auxillary Winding.
- $c5\{J\}$ = % of turns per pole in each coils in Aux-wdg.

- c6[J] = Total Series turns in each coils for Aux-wdg.
- ✓ d = Diameter of the stator core.
- ✓ d1 = Nearest Standard Diameter of the Stator Core.
- ✓ dr = Outer Diameter of the Rotor.
- ✓ do = Outer Diameter of the Stator Core.
- ✓ dcr = Depth of the Rotor Core.
- ✓ dem = Mean Diameter of the End Ring.
- ✓ dob = Depth of the Bar.
- ✓ doe = Depth of the End Ring.
- ✓ dsc = Depth of the Stator Core.
- ✓ dss = Depth of the Stator Slot.
- ✓ ds1 = Calculated depth of stator slot
- ✓ deri = Inner Diameter of the End Ring.
- ✓ dero = Outer Diameter of the End Ring.

- ✓ dors = Depth of the Rotor Slot.
- ✓ e = Induced EMF in the machine.
- ✓ eff = Input (or) Required Efficiency.
- err = End ring resistance at 75 degree.
- effl = Desinged Efficiency.
- ermw = Effective Rotor Resistance in terms of Main Wdg.
- fdt = Maximum flux density in the Stator teeth.
- ✓ FP fpp = Flux per Pole.
- fwl = Total Friction and Windage Losses.
- ✓ hp = Horse of the Machine.
- ✓ i = Rated Current of the Machine.
- ✓ ib = Rotor Bar Current.
- ✓ id = Current Density(Amps/sq.mm).
- ✓ ie = End Ring Current.

il	=	No load Losses.
im	=	Magnetising current.
io	=	No load current.
ip	=	Input of the machine.
ics	=	Inner cylindrical surface of stator.
idr	=	Inner diameter of the rotor.
isc	=	Short Ckt Current.
kl	=	Leakage Factor.
kg	=	Gap Contraction factor.
✓ kwa	=	Winding Factor of the Auxillary Winding.
✓ kwm	=	Winding Factor of the Main Winding.
kcs	=	Carter's Co-efficient Factor of Stator Slots.
kcr	=	Carter's Co-efficient Factor of Rotor Slots.
kgss	=	Gap Contraction Factor of Stator Slots.

kg_{sr} = Gap Contraction Factor of Rotor Slots.

✓ l = Length of the Stator Core.

l_l = Nett Iron Length of Stator Core.

l_b = Length of each bar in the Rotor.

✓ l_g = Length of the Air Gap.

✓ l_{ip} = Slot Opening of the Stator Slot.

l_{dos} = Loss dissipated from outer surface

l_{dis} = Loss dissipated from inner surface.

l_{ded} = Loss dissipated from outer surface.

l_{ipr} = Slot Opening of the Rotor Slot.

l_{rca} = Locked Rotor Current in Auxillary Wdg.

l_{rcm} = Locked Rotor Current in Main Winding.

✓ l_{mtm} = Length of mean turn of Main Winding.

✓ l_{mts} = Length of mean turn of Starting Winding.

lmt[j] = Length of mean turn of each coil for Main Winding.

lmta[j]= Length of mean turn of each coil of Starting Wdg.

m = $d^2 \times l$ value.

max = Total Conductor Area in the Stator Windings.

n = Synchronous Speed in RPM.

✓ ns = Synchronous Speed in RPS.

ntm = Calculated number of series turns in Main Wdg.

✓ neck = Neck of the Stator Slots.

nlpf = No-Load Power Factor.

✓ neckr = Neck of the Rotor Slots.

op = Motor Output in Watts.

opl = Desinged motor output.

ocs = Outer cylindrical surface of stator.

p = Number of poles of the machine.

- p_l = Length to diameter value.
- pf = Input Power Factor.
- pac = Nearest Standard Capacitor.
- q = Iron Loss per Kilogram in tooth section.
- q_l = Iron Loss per Kilogram in core section.
- r = Average coil span.
- r_l = Sum of the coil span.
- r_v = Rated Voltage of the machine.
- r_r = Resistance of the Rotor at 75 Degrees.
- r_{r1} = Resistance of Rotor referred to Main Wdg at 75 Deg.
- r_{r2} = Resistance of Rotor referred to Main Wdg at 20 Deg.
- r_{s1} = Resistance of Main Winding at 75 Degrees.
- r_{s2} = Resistance of Main Winding at 20 Degrees.
- r_{sp} = Rotor slot pitch.

- ✓ rsw = Rotor slot width.
- sa = Area of the Stator Slots.
- sd = Shaft diameter.
- ✓ sr = Number of Rotor slots.
- ✓ ss = Number of Stator slots.
- st = Starting torque of the machine.
- ✓ ssp = Stator slot pitch.
- ✓ spp = Stator Pole Pitch.
- spr = Specific permeance of rotor slots.
- sps = Specific permeance of stator slots.
- spz = Specific permeance of zig-zag leakage.
- scr1 = Lower semi-circle diameter of stator slots.
- scr2 = Upper semi-circle diameter of stator slots.
- scpf = Short circuit Power factor

ta	=	Calculated total series turns in Aux-Wdg.
tl	=	Total loss in the Machine.
tr	=	Temperature Rise in machine.
tcm	=	Total Copper Loss in the Machine.
tid	=	Total Loss dissipation.
til	=	Total Iron Loss in the Stator.
tim	=	Total Iron Loss in the Machine.
tnl	=	Total no load losses.
tpm	=	Total series turns per pole.
tta	=	Total series turns in Auxillary Wdg.
ttm	=	Actual series turns in the Main Wdg.
tou	=	Pole Pitch of the Machine.
tsc	=	Total Starting current.
tts	=	Total tooth section in Stator.

- ✓ t_{csm} = Total Copper Section in Main Wdg.
- ✓ t_{csr} = Total Conductor Section in Rotor Bar.
- $trrs$ = Total rotor resistance in terms of starting wdg
- $trsw$ = Total resistance in starting winding
- $tlrs$ = Total leakage reactance in starting winding.
- $tlds$ = Total Loss to be dissipated by stator.
- $trmw$ = Total resistance in terms of main winding
- $tppa$ = Total series turns per pole in Auxillary wdg
- $tirs$ = Total impedance of rotor at Standstill.
- $trst$ = Total rotor resistance in terms of Starting wdg.
- va = Peripheral speed of Machine.
- vae = Velocity of air at two End surface.
- ✓ wtr = Width of the Rotor Teeth.
- wgs = Weight of Stator Core.

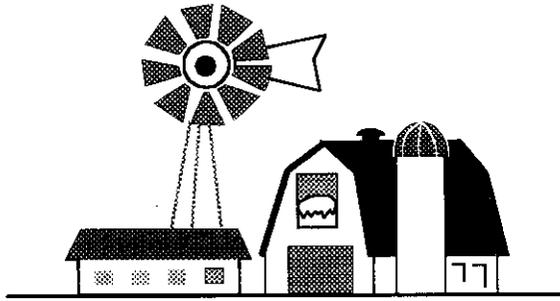
- wgt = Weight of Stator teeth.
- wrt = Width of the Rotor teeth at Slot opening.
- ✓ word = Width of the Rotor Bar.
- ✓ wors = Width of the Rotor Slot.
- wsst = Width of the Stator teeth at Slot opening.
- x = Length to pole pitch ratio.
- xc = Calculated Capacitive Reactance.
- xs = Slot Leakage Reactance.
- xm = Magnetic Reactance.
- xz = Zig-Zag Leakage Reactance.
- xc1 = Standard Capacitive Reactance.
- xsl = Skew Leakage Reactance.
- xlm = Total Leakage Reactance.
- xom = Open Circuit Reactance.

z_1 = Diameter of the Main winding.

z_2 = Diameter of the Auxillary winding.

z_a = Impedance of Starting Wdg with Capacitor in series.

z_m = Impedance of Main Winding(locked).



ENERGY

Efficient MOTOR

C H A P T E R -7

ENERGY EFFICIENT MOTORS

7.1 ENERGY EFFICIENT DRIVES

Energy conservation is the foremost task of the day and efforts are being made all over to bring down the power consumption in the wake of present supply/demand position of power with a high cost added to it. Efficient generation, transmission, and distribution are no doubt the important factors to increase the available power. But for a given power, it's efficient usage plays the most important role.

A conventional motor made these days has a high output to weight ratio and reduced material cost thereby compromising on factors like efficiency and power factor. It is interesting to note that the cost of operating a motor is 4 to 6 times it's original cost. Also the general trend is to employ over sized motor and provide a safe service factor.

Under reduced load conditions controlled energy saving plays a significant role. In addition to providing a soft starting, it reduces the power absorbed by the motor at low loads and improves the power factor and hence this controller in conjunction with an energy efficient motor (which has a higher efficiency even at low loads) would give tremendous energy saving.

An incremental improvement in efficiency of these motors can have significant effect in reducing the energy bills.

7.2 ENERGY EFFICIENT MOTOR

Though electric motors are one of the most efficient means of converting electrical energy into mechanical, yet some losses do occur in this conversion process. The losses that occur are comprised of five components namely:

	LOSS COMPONENT	APPROX %
1.	Primary	40
2.	Iron	25
3.	Secondary	20
4.	Friction & Windage	10
5.	Stray load	05
	Total	100

The primary/secondary losses are due to the passage of the current in the windings. Increasing the cross-sectional area of copper/aluminium to reduce the resistance is the only effective means to reduce the losses. Of course, this will add to the

copper/aluminium weight and subsequently the cost of motor will increase. Also reduction in secondary resistance will reduce the starting torque of the motor.

The iron losses occur in the machine due to the combined effect of hysteresis and eddy current losses. These losses are a result of the magnetic field in the machine core which oscillate at line frequency. These losses can be reduced by reducing the flux density and this can be achieved by adding length to the stator core. Further reduction is achieved by using a better grade of steel and reducing the thickness of lamination. Increasing silicon watt loss per kg and reducing the thickness effects a reduction in eddy current losses.

The friction and windage losses are those associated with the cooling fans and rotor fins. The stray load losses are the most elusive and are caused by imperfections in and around the air gap length. This includes stator/rotor slotting, nonuniform air gaps, rough surfaces and bulging of stampings. A combination of design and careful manufacturing practices can reduce the stray load losses.

ENERGY FACTORS IN THE SELECTION OF ELECTRIC MOTORS

Choosing a motor most suitable for a particular application is based on many factors, including the requirements of the driven equipments, service conditions, motor efficiency and motor power factor. Good energy management is the application of the motor and its components that results in the least consumption energy.

7.3 MOTOR EFFICIENCY

Motor efficiency is a measure of the effectiveness with which a motor converts electrical energy into mechanical energy to perform work. It is input or output in terms of electrical power, watts input to watts output plus watt losses. Watt losses are electrical energy that is not converted into useful mechanical energy, but is converted into heat and dissipated through the motor frame. Watt losses are electrical losses, friction and windage.

Thus, motor efficiency is a function of watt losses. The lower the watt losses of a given motor, the higher the efficiency. Even in higher horse power motors that typically have higher efficiency percentage than lower horse power motors, watt losses are significant and cannot be ignored.

A.C. motors have five components of watt losses that contribute to motor efficiency :

1. Iron loss.
2. Stator copper loss
3. Rotor copper loss.
4. Friction and windage loss.
5. Stray load loss.

The losses are associated with three common elements of motor construction :

1. losses in the circuits which carry current(I^2R)
2. losses in the circuits that carry magnetic flux (Iron)
3. losses in those parts which help support and ventilate or cool motor parts.(friction and windage loss).
4. Stray losses can occur in any or all of the above elements.

Improvements in motor efficiency is the result of reducing watt losses.Simply stated :

REDUCED LOSSES - IMPROVED EFFICIENCY.

In terms of existing state of electric motor technology, a reduction in watt losses can be in various ways.

1. Increasing the cross sectional area of the winding using a greater number of conductors, or large conductor crosssection, increasing slot fill and reducing stator and rotor copper losses ;
2. Adding more core steel to reduce operating flux density ;

3. Utilize low loss, core steel for stator and rotor laminations to improve magnetic properties ;
4. Using low loss motor cooling fans to reduce winding losses;
5. Using trial and error method find out good combination of core diameter and core length to have a reduced loss of machine.

All of these changes to reduce watt loss are also possible with existing motor design and manufacturing technology. They do, however, require additional materials and for the use of higher quality material resulting in increased motor cost.

7.4 FACTORS AFFECTING EFFICIENCY OF INDUCTION MOTORS

VOLTAGE IMBALANCE : A voltage unbalance of 3.5% can increase the motor losses by 20%.

LOAD : Motor efficiency changes as the load changes. Operation at loads substantially different from rate load may also result in a change in efficiency.

HORSE POWER : The efficiency of the motor operating at rated load increases as the motor horsepower rating increases. High horsepower motors are more efficient.

DUTY CYCLE: Motors that operate continuously and for long periods of time provide energy reducing opportunities. A small change in motor efficiency will make a significant change in total energy consumed. Examples: PUMPS, FANS, BLOWERS.

SLIP: Slip is measure of the losses in the rotor winding of poly phase induction motor. Therefore the higher the slip, the lower the efficiency. For applications involving pulsating intertia pumping the overall efficiency of motors with high slip may be higher than of motors having a slip of less than 5 percentage.

7.5 **ADVANTAGES OF ENERGY EFFICIENT MOTOR**

The advantages of an energy efficient motor are:

1. Reduced power consumption.
2. Lower temperature rise and comparatively increased service life.
3. Broader band of constant efficiency thereby operating at light loads without appreciable drop in efficiency.

OUTPUT PARAMETERS OF 0.25 H.P SINGLE PHASE INDUCTION MOTOR

$B_m=0.52$ Wb/m², $a_c=9000$, $N=1440$ rpm, $f=50$ Hz)

STATOR

THE SYNCHRONOUS SPEED OF THE MACHINE :1500 RPM
THE NUMBER OF THE POLES OF THE MACHINE :4.000000

THE STATOR DIMENSIONS

THE CALCULATED DIAMETER OF THE MACHINE :0.062124 METRES
THE NEAREST STANDRISED DIAMETER IS :0.058000 METRES
THE CORE LENGTH OF THE MACHINE :0.073218 METRES
THE POLE PITCH OF THE MACHINE :0.048812 METRES
THE NET IRON LENGTH OF THE MACHINE :0.069557 METRES
THE FLUX PER POLE : 0.001858 WEBERS
THE OUTER DIAMETER OF THE CORE :0.105000 METRES

THE STATOR SLOT PARAMETRES

THE DIAMETER OF THE BOTTOM SEMI CIRCULAR SLOT :0.005064 METRES
THE DIAMETER OF THE UPPER SEMI CIRCULAR SLOT :0.006414 METRES
THE NECK OF THE STATOR SLOT :0.778700 MM
THE SLOT OPENING :2.228000 MM
THE TOOTH WIDTH OF STATOR :0.003071 METRES
THE TOTAL TOOTH SECTION :0.005127
THE AVERAGE SLOT WIDTH :0.005739 METRES
THE DEPTH OF THE STATOR CORE :0.011133 METRES
THE DEPTH OF THE STATOR SLOT:0.010305 METRES

THE MAIN WINDING PARAMETERS

THE WINDING FACTOR OF MAIN WINDING:0.776457
THE INDUCED EMF IS:218.500000
THE NUMBER OF TURNS IN SERIES PER POLE :171
THE TOTAL TURNS IN SERIES FOR THE MAIN WINDING :684
THE RATED CURRENT OF THE MACHINE :1.249151 AMPS
THE AREA OF THE MAIN WINDING IS :1.665534e-07 SQ.METRES
THE DIAMETER OF THE CONDUCTOR :0.000460 METRES
THIS DIAMETER IS NOT CONSIDER THE ENAMEL THICKNESS
THE TURNS OF THE COIL (6,1)=86
THE TURNS OF THE COIL (5,2)=63
THE TURNS OF THE COIL (4,3)=23
THE LENGTH OF MEAN TURN OF COIL (6,1)=0.273187
THE LENGTH OF MEAN TURN OF COIL (5,2)=0.222487
THE LENGTH OF MEAN TURN OF COIL (4,3)=0.171786
THE LENGTH OF THE MEAN TURN OF THE MAIN WINDING =0.242467

STARTING WINDING PARAMETERS

THE AREA OF THE STARTING WINDING IS :7.570610e-08 SQ.METRES
THE DIAMETER OF THE STARTING WINDING IS:0.000310 METRES
THIS DIAMETER IS NOT CONSIDER THE ENAMEL THICKNESS
THE WINDING FACTOR OF STARTING WINDING:0.845185
THE TURNS PER POLE IN STARTING WINDING:241
THE TOTAL TURNS IN STARTING WINDING:964
THE TURNS OF THE COIL(10,4)=102
THE TURNS OF THE COIL(9,5)=89
THE TURNS OF THE COIL(8,6)=51
THE LENGTH OF MEAN TURN OF COIL(10,4)=0.298538
THE LENGTH OF MEAN TURN OF COIL(9,5)=0.247837
THE LENGTH OF MEAN TURN OF COIL(8,6)=0.197136
THE LENGTH OF MEAN TURN OF STARTING WINDING :0.259594
THE LENGTH OF AIR GAP OF THE MACHINE IS :0.334886 MILLIMETRES

ROTOR DIMENSIONS

THE ALUMINIUM ROTOR PARAMETERS ARE :

NUMBER OF THE ROTOR SLOTS IS:18.000000
OUTER DIAMETER OF THE ROTOR IS:0.061454 METRES
WIDTH OF THE ROTOR SLOT IS:0.006768 METRES
DEPTH OF THE ROTOR SLOT IS:0.002623 METRES
TOTAL CONDUCTOR AREA IN ROTOR IS :0.000251 M²
WIDTH OF THE BAR IS :0.006768 METRES
DEPTH OF THE BAR IS :0.001556 METRES
AREA OF THE BAR IS :0.000011 SQ.METRES
LENGTH OF EACH BAR IS :0.073668 METRES
WIDTH OF THE ROTOR TOOTH IS :0.003958 METRES
DEPTH OF THE ROTOR CORE IS : 0.008349 METRES
NECK OF THE ROTOR SLOT IS : 1.066800 MM
GAP OF THE ROTOR SLOT IS : 1.000000 MM
INNER DIAMETER OF ROTOR IS: 0.039510 METRES
BAR CURRENT IS:58.970650 AMPS

END RING

DEPTH OF THE END RING IS:0.010000 METRES
THICKNESS OF THE END RING IS :0.012005 METRES
OUTER DIAMETER OF THE END RING IS:0.058454 METRES
INNER DIAMETER OF THE END RING IS:0.038454 METRES
MEAN DIAMETER OF THE END RING IS:0.048454 METRES
AREA OF THE EACH END RING IS :0.000120 M²
END RING CURRENT IS :84.469231 AMPS
DIAMETER OF THE SHAFT IS:0.039510 METRES

RESISTANCE AND REACTANCE OF WINDINGS

GAP CONTRACTION FACTOR IS GOT TO BE :1.300626
RESISTANCE OF MAIN WINDING AT 20C IS :16.927927 OHMS
RESISTANCE OF MAIN WINDING AT 75C IS :20.910969 OHMS
RESISTANCE OF ROTOR AT 75C IS :0.000147 OHMS
RESISTANCE OF EACH END RING AT 75C IS :0.000008 OHMS
RESISTANCE OF ROTOR REFERRED TO MAIN WINDING AT 20C IS :15.502359 OH
RESISTANCE OF ROTOR REFERRED TO MAIN WINDING AT 75C IS :19.149973 OH
SLOT LEKEAGE REACTANCE IN TERMS OF MAIN WINDING IS :11.724553 OHMS
ZIGZAG LEAKAGE REACTANCE IS :7.090288 OHMS
OVERHANG REACTANCE IS :16.846352 OHMS
MAGNETIZING REACTANCE IS :159.030350 OHMS
SKEW LEAKAGE REACTANCE IS :3.454373 OHMS
TOTAL LEAKAGE REACTANCE REFERRED TO MAIN WINDING IS :39.115566 OHMS
OPEN CKT REACTANCE IS :178.588135 OHMS
LEAKAGE FACTOR IS :0.883727

CAPACITOR DESIGN VALUES

CALCULATED VALUE OF CAPACITIVE REACTANCE :91.837433 OHMS
CALCULATED CAPACITANCE:34.646202 MICROF
NEAREST STANDARD VALUE IS :25.000000 MICROF *
STANDARD VALUE OF CAPACITIVE REACTANCE IS :127.272728 OHMS

MACHINE LOSSES

MAXIMUM FLUX IN THE STATOR TEETH IS:2.278572 WEBERS
TOTAL IRON LOSS IN THE STATOR :7.535411 WATTS
TOTAL IRON LOSS IN THE ROTOR :7.535411 WATTS
TOTAL COPPER LOSS IN THE STATOR:32.629009 WATTS
TOTAL COPPER LOSS IN THE ROTOR:9.197342 WATTS
TOTAL COPPER LOSS IN THE END RING IS :0.120953 WATTS
TOTAL IRON LOSS IN THE MACHINE:15.070823 WATTS
TOTAL COPPER LOSS IN THE MACHINE:41.947304 WATTS
TOTAL FRICTION AND WINDAGE LOSSES IN THE MACHINE:0.919375 WATTS
TOTAL LOSSES IN THE MACHINE IS:57.937500 WATTS

MACHINE CHARACTERSTIC DETAILS

LINE CURRENT OF START WDG WITH CAPACITOR IN SERIES (A) IS : 11.52454 AMPS
LINE CURRENT IN MAIN WDG IS : 4.352454 AMPS
LINE CURRENT IN AUXI WDG IS : 2.173576 AMPS
LINE CURRENT IS : 6.526030 AMPS
LINE TORQUE IS : 1.979290 NEWTON METRE
LINE CURRENT IS : 4.258329 AMPS
LINE POWER FACTOR IS : 0.746933
LINE IS GIVEN BY : 214.597229
CALCULATED OUTPUT OF THE MACHINE IS : 1106.609779
CALCULATED EFFICIENCY OF MACHINE IS : 73.60174,

NO LOAD CHARACTERSTICS

NO LOAD CURRENT IS : 2.317287 AMPS
NO LOAD LOSS IS : 15.990198 WATTS
NO LOAD CURRENT IS : 2.318330 AMPS
NO LOAD POWER FACTOR IS : 0.030002

MMF VALUES OF MACHINE

MMF IN THE AIR GAP IS : 284.474365 AMPS
MMF IN THE STATOR TEETH IS : 350.383514 AMPS
MMF IN THE ROTOR TEETH IS : 49.829044 AMPS
MMF IN THE STATOR CORE IS : 11.052895 AMPS
MMF IN THE ROTOR CORE IS : 24.812551 AMPS
TOTAL MMF IN THE MACHINE IS : 720.552368 AMPS

TEMPERATURE RISE

LOSS DISSIPATED BY STATOR SURFACE IS : 27.741358 WATTS
LOSS DISSIPATION : 1.001567 WATTS
TEMPERATURE RISE OF MACHINE IS : 27.193780 DEGREE
TEMPERATURE RISE IS WITHIN THE LIMIT
ALL MACHINE ALL PARAMETER VALUES MOTOR*****

*** PLEASE DO NOT USE THIS SOFTWARE *****

PARAMETERS OF 0.5 H.P SINGLE PHASE INDUCTION MOTOR

$B_m = 0.52$ wb/m², $\omega_c = 9000$, $N = 1440$ rpm, $f = 50$ Hz)

STATOR

SYNCHRONOUS SPEED OF THE MACHINE : 1500 RPM
THE NUMBER OF THE POLES OF THE MACHINE : 4.000000

THE STATOR DIMENSIONS

REGULATED DIAMETER OF THE MACHINE : 0.078272 METRES
NEAREST STANDARDISED DIAMETER IS : 0.077900 METRES
CORE LENGTH OF THE MACHINE : 0.092249 METRES
SOLE PITCH OF THE MACHINE : 0.061499 METRES
NET IRON LENGTH OF THE MACHINE : 0.087636 METRES
FLUX PER POLE : 0.002950 WEBERS
CORE DIAMETER OF THE CORE : 0.139000 METRES

THE STATOR SLOT PARAMETERS

DIAMETER OF THE BOTTOM SEMI CIRCULAR SLOT : 0.004785 METRES
DIAMETER OF THE UPPER SEMI CIRCULAR SLOT : 0.006390 METRES
DEPTH OF THE STATOR SLOT : 0.778700 MM
SLOT OPENING : 2.228000 MM
TOOTH WIDTH OF STATOR : 0.002902 METRES
TOOTH WIDTH SECTION : 0.008138
TOOTH SLOT WIDTH : 0.005588 METRES
DEPTH OF THE STATOR CORE : 0.014026 METRES
DEPTH OF THE STATOR SLOT : 0.016338 METRES

MAIN WINDING PARAMETERS

WINDING FACTOR OF MAIN WINDING:0.780351
AIR GAP OF THE MACHINE IS:218.500000
NUMBER OF TURNS IN SERIES PER POLE :107
NUMBER OF TURNS IN SERIES FOR THE MAIN WINDING :428
SPEED OF THE MACHINE :1498700 RPM
AREA OF THE MAIN WINDING IS :3.831069e-07 SQ.METRES
DIAMETER OF THE CONDUCTOR :0.000651 METRES
DIAMETER IS NOT CONSIDER THE ENAMEL THICKNESS=****
TURNS OF THE COIL (8,1)=11
TURNS OF THE COIL (7,2)=35
TURNS OF THE COIL (6,3)=74
TURNS OF THE COIL (5,4)=77
LENGTH OF MEAN TURN OF COIL (8,1)=0.358342
LENGTH OF MEAN TURN OF COIL (7,2)=0.308672
LENGTH OF MEAN TURN OF COIL (6,3)=0.259002
LENGTH OF MEAN TURN OF COIL (5,4)=0.209332
LENGTH OF THE MEAN TURN OF THE MAIN WINDING =0.313978

STARTING WINDING PARAMETERS

AREA OF THE STARTING WINDING IS :1.514122e-07 SQ.METRES
DIAMETER OF THE STARTING WINDING IS:0.000439 METRES
DIAMETER IS NOT CONSIDER THE ENAMEL THICKNESS=****
WINDING FACTOR OF STARTING WINDING:0.829425
AIR GAP PER POLE IN STARTING WINDING:155
NUMBER OF TURNS IN STARTING WINDING:620
TURNS OF THE COIL(13,5)=52
TURNS OF THE COIL(12,6)=48
TURNS OF THE COIL(11,7)=37
TURNS OF THE COIL(10,8)=20
LENGTH OF MEAN TURN OF COIL(13,5)=0.283177
LENGTH OF MEAN TURN OF COIL(12,6)=0.333507
LENGTH OF MEAN TURN OF COIL(11,7)=0.283837
LENGTH OF MEAN TURN OF COIL(10,8)=0.234167
LENGTH OF MEAN TURN OF STARTING WINDING :0.329779
AIR GAP OF AIR GAP OF THE MACHINE IS :0.389946 METRES

ROTOR DIMENSIONS

ALUMINIUM ROTOR PARAMETERS ARE :

NUMBER OF THE ROTOR SLOTS IS:44.000000
OUTER DIAMETER OF THE ROTOR IS:0.077532 METRES
WIDTH OF THE ROTOR SLOT IS:0.003496 METRES
DEPTH OF THE ROTOR SLOT IS:0.019224 METRES
TOTAL CONDUCTOR AREA IN ROTOR IS :0.000314 M²
WIDTH OF THE BAR IS :0.003496 METRES
DEPTH OF THE BAR IS :0.018157 METRES
AREA OF THE BAR IS :0.000063 SQ.METRES
LENGTH OF EACH BAR IS :0.092568 METRES
WIDTH OF THE ROTOR TOOTH IS :0.002040 METRES
WIDTH OF THE ROTOR CORE IS : 0.010520 METRES
WIDTH OF THE ROTOR SLOT IS : 1.066800 MM
WIDTH OF THE ROTOR SLOT IS : 1.000000 MM
INNER DIAMETER OF ROTOR IS: 0.018045 METRES
ROTOR CURRENT IS:30.342520 AMPS

END RING

DEPTH OF THE END RING IS:0.010000 METRES
THICKNESS OF THE END RING IS :0.012509 METRES
OUTER DIAMETER OF THE END RING IS:0.074532 METRES
INNER DIAMETER OF THE END RING IS:0.054532 METRES
MEAN DIAMETER OF THE END RING IS:0.064532 METRES
AREA OF THE EACH END RING IS :0.000125 M²
END RING CURRENT IS :106.241562 AMPS
DIAMETER OF THE SHAFT IS:0.018045 METRES

RESISTANCE AND REACTANCE OF WINDINGS

POWER FACTOR IS GOT TO BE :1.342239
RESISTANCE OF MAIN WINDING AT 200 IS :6.955136 OHMS
RESISTANCE OF MAIN WINDING AT 750 IS :8.471851 OHMS
RESISTANCE OF ROTOR AT 750 IS :0.000031 OHMS
RESISTANCE OF EACH END RING AT 750 IS :0.000011 OHMS
RESISTANCE OF ROTOR REFERRED TO MAIN WINDING AT 200 IS :0.000011 OHMS
RESISTANCE OF ROTOR REFERRED TO MAIN WINDING AT 750 IS :0.000031 OHMS
LEAKAGE REACTANCE IN TERMS OF MAIN WINDING IS :3.24041 OHMS
LEAKAGE REACTANCE IS :0.872754 OHMS
PERMANENT REACTANCE IS :8.702767 OHMS
MAGNETIZING REACTANCE IS :87.574959 OHMS
LEAKAGE REACTANCE IS :1.070019 OHMS
TOTAL LEAKAGE REACTANCE REFERRED TO MAIN WINDING IS :13.805401 OHMS
MOTOR REACTANCE IS :94.517952 OHMS
POWER FACTOR IS :0.923627

CAPACITOR DESIGN VALUES

REGULATED VALUE OF CAPACITIVE REACTANCE :33.188747 OHMS
REGULATED CAPACITANCE:95.870392 MICROF
STANDARD VALUE IS :25.000000 MICROF
STANDARD VALUE OF CAPACITIVE REACTANCE IS :127.272728 OHMS

MACHINE LOSSES

MAXIMUM FLUX IN THE STATOR TEETH IS:2.278572 WEBERS
TOTAL IRON LOSS IN THE STATOR :16.141581 WATTS
TOTAL IRON LOSS IN THE ROTOR :16.141581 WATTS
TOTAL COPPER LOSS IN THE STATOR:52.877148 WATTS
TOTAL COPPER LOSS IN THE ROTOR:1.240729 WATTS
TOTAL COPPER LOSS IN THE END RING IS :0.244557 WATTS
TOTAL IRON LOSS IN THE MACHINE:32.283161 WATTS
TOTAL COPPER LOSS IN THE MACHINE:54.362434 WATTS
TOTAL FRICTION AND WINDAGE LOSSES IN THE MACHINE:1.838750 WATTS
TOTAL LOSSES IN THE MACHINE IS:88.484344 WATTS

MACHINE CHARACTERISTIC DETAILS

1. CAPACITANCE OF START WDG WITH CAPACITOR IN SERIES IS:98.484285 μF
2. NO LOAD ROTOR CURRENT IN MAIN WDG IS:14.432155 AMPS
3. NO LOAD ROTOR CURRENT IN AUX1 WDG IS:2.335853 AMP
4. NO LOAD STARTING CURRENT IS:16.769017 AMPS
5. NO LOAD STARTING TORQUE IS:0.273977 NEWTON-METER
6. NO LOAD IFT CURRENT IS:14.437675 AMPS
7. NO LOAD IFT POWER FACTOR IS:0.593966
8. NO LOAD IFT IS GIVEN BY:341.298279
9. ESTIMATED OUTPUT OF THE MACHINE IS :252.913934
10. ESTIMATED EFFICIENCY OF MACHINE IS:74.074182

NO LOAD CHARACTERISTICS

1. NO LOAD IFT CURRENT IS :4.676652 AMPS
2. NO LOAD LOSS IS :34.121910 WATTS
3. NO LOAD CURRENT IS :4.679005 AMPS
4. NO LOAD POWER FACTOR IS :0.031723

MMF VALUES OF MACHINE

1. MMF IN THE AIR GAP IS:324.310974 AMPS
2. MMF IN THE STATOR TEETH IS:555.496582 AMPS
3. MMF IN THE ROTOR TEETH IS :8.650649 AMPS
4. MMF IN THE STATOR CORE IS :14.715676 AMPS
5. MMF IN THE ROTOR CORE IS :11.332358 AMPS
6. TOTAL MMF IN THE MACHINE IS:914.506226 AMPS

TEMPERATURE RISE

1. LOSS DISSIPATED BY STATOR SURFACE IS :47.212841 WATTS
2. LOSS DISSIPATION :1.694069 WATTS
3. TEMPERATURE RISE OF MACHINE IS :27.869497 DEGREE
4. TEMPERATURE RISE IS WITHIN THE LIMIT
5. COMPLETE ALL PARAMETER VALUES MOTOR*****

6. THANK YOU FOR USING SOFTWARE*****

PARAMETERS OF 1.0 H.P SINGLE PHASE INDUCTION MOTOR

$B = 0.52 \text{ Wb/m}^2$, $\rho_c = 9000$, $N = 1440 \text{ mm}$, $f = 50 \text{ Hz}$

STATOR

SYNCHRONOUS SPEED OF THE MACHINE : 1500 RPM
POLES OF THE POLES OF THE MACHINE : 4.000000

THE STATOR DIMENSIONS

OUTER DIAMETER OF THE MACHINE : 0.094819 METRES
INNER DIAMETER OF THE MACHINE IS : 0.093500 METRES
CORE LENGTH OF THE MACHINE : 0.111751 METRES
POLE PITCH OF THE MACHINE : 0.074500 METRES
CORE IRON LENGTH OF THE MACHINE : 0.106164 METRES
FLUX PER POLE : 0.004329 WEBERS
INNER DIAMETER OF THE CORE : 0.156000 METRES

THE STATOR SLOT PARAMETERS

DIAMETER OF THE BOTTOM SEMI CIRCULAR SLOT : 0.005797 METRES
DIAMETER OF THE UPPER SEMI CIRCULAR SLOT : 0.007133 METRES
DEPTH OF THE STATOR SLOT : 0.778700 MM
SLOT OPENING : 2.228000 MM
SLOT WIDTH OF STATOR : 0.003515 METRES
SLOT TOOTH SECTION : 0.011943
AVERAGE SLOT WIDTH : 0.006465 METRES
WIDTH OF THE STATOR CORE : 0.016991 METRES
DEPTH OF THE STATOR SLOT : 0.013599 METRES

THE MAIN WINDING PARAMETERS

WINDING FACTOR OF MAIN WINDING:0.780361
EMF INDUCED EMF IS:218.500000
NUMBER OF TURNS IN SERIES PER POLE 473
TOTAL TURNS IN SERIES FOR THE MAIN WINDING :292
RATED CURRENT OF THE MACHINE :4.441425 AMPS
AREA OF THE MAIN WINDING IS :5.921900e-07 SQ.METRES
DIAMETER OF THE CONDUCTOR :0.000868 METRES
THIS DIAMETER IS NOT CONSIDER THE ENAMEL THICKNESS
TURNS OF THE COIL (8,1)=28
TURNS OF THE COIL (7,2)=24
TURNS OF THE COIL (6,3)=16
TURNS OF THE COIL (5,4)=6
LENGTH OF MEAN TURN OF COIL (8,1)=0.422721
LENGTH OF MEAN TURN OF COIL (7,2)=0.365801
LENGTH OF MEAN TURN OF COIL (6,3)=0.308882
LENGTH OF MEAN TURN OF COIL (5,4)=0.251962
LENGTH OF THE MEAN TURN OF THE MAIN WINDING =0.370812

STARTING WINDING PARAMETERS

AREA OF THE STARTING WINDING IS :2.691773e-07 SQ.METRES
DIAMETER OF THE STARTING WINDING IS:0.000585 METRES
THIS DIAMETER IS NOT CONSIDER THE ENAMEL THICKNESS
WINDING FACTOR OF STARTING WINDING:0.829425
TURNS PER POLE IN STARTING WINDING:106
TOTAL TURNS IN STARTING WINDING:424
TURNS OF THE COIL(13,5)=36
TURNS OF THE COIL(12,6)=33
TURNS OF THE COIL(11,7)=25
TURNS OF THE COIL(10,8)=14
LENGTH OF MEAN TURN OF COIL(13,5)=0.451181
LENGTH OF MEAN TURN OF COIL(12,6)=0.394261
LENGTH OF MEAN TURN OF COIL(11,7)=0.337342
LENGTH OF MEAN TURN OF COIL(10,8)=0.280422
LENGTH OF MEAN TURN OF STARTING WINDING :0.392571
LENGTH OF AIR GAP OF THE MACHINE IS :0.405673 MILLIMETRES

ROTOR DIMENSIONS

THE ALUMINIUM ROTOR PARAMETERS ARE :

NUMBER OF THE ROTOR SLOTS IS:44.000000
OUTER DIAMETER OF THE ROTOR IS:0.094007 METRES
WIDTH OF THE ROTOR SLOT IS:0.004241 METRES
DEPTH OF THE ROTOR SLOT IS:0.029260 METRES
TOTAL CONDUCTOR AREA IN ROTOR IS :0.000380 M²
WIDTH OF THE BAR IS :0.004241 METRES
LENGTH OF THE BAR IS :0.028193 METRES
AREA OF THE BAR IS :0.000120 SQ.METRES
LENGTH OF EACH BAR IS :0.112139 METRES
WIDTH OF THE ROTOR TOOTH IS :0.002471 METRES
DEPTH OF THE ROTOR CORE IS : 0.012744 METRES
RADIUS OF THE ROTOR SLOT IS : 1.066800 MM
WIDTH OF THE ROTOR SLOT IS : 1.000000 MM
INNER DIAMETER OF ROTOR IS: 0.010001 METRES
BAR CURRENT IS:36.801731 AMPS

END RING

DEPTH OF THE END RING IS:0.010000 METRES
THICKNESS OF THE END RING IS :0.013043 METRES
OUTER DIAMETER OF THE END RING IS:0.091007 METRES
INNER DIAMETER OF THE END RING IS:0.071007 METRES
MEAN DIAMETER OF THE END RING IS:0.081007 METRES
AREA OF THE EACH END RING IS :0.000130 M²
END RING CURRENT IS :128.857910 AMPS
DIAMETER OF THE SHAFT IS:0.010001 METRES

RESISTANCE AND REACTANCE OF WINDINGS

CONTRACTION FACTOR IS GOT TO BE :1.283335
RESISTANCE OF MAIN WINDING AT 200 IS :3.108314 OHMS
RESISTANCE OF MAIN WINDING AT 750 IS :3.839681 OHMS
RESISTANCE OF ROTOR AT 750 IS :0.000020 OHMS
RESISTANCE OF EACH END RING AT 750 IS :0.000013 OHMS
RESISTANCE OF ROTOR REFERRED TO MAIN WINDING AT 200 IS :11.37953 OHMS
RESISTANCE OF ROTOR REFERRED TO MAIN WINDING AT 750 IS :16.392603 OHMS
LEAKAGE REACTANCE IN TERMS OF MAIN WINDING IS :1.855746 OHMS
LEAKAGE REACTANCE IS :0.677260 OHMS
OVERHANG REACTANCE IS :4.641968 OHMS
MAGNETIZING REACTANCE IS :57.004620 OHMS
LEAKAGE REACTANCE IS :0.696501 OHMS
TOTAL LEAKAGE REACTANCE REFERRED TO MAIN WINDING IS :17.844413 OHMS
LEAKAGE REACTANCE IS :60.939358 OHMS
LEAKAGE FACTOR IS :0.933201

CAPACITOR DESIGN VALUES

CALCULATED VALUE OF CAPACITIVE REACTANCE :19.084885 OHMS
CALCULATED CAPACITANCE:166.719284 MICROF
NEAREST STANDARD VALUE IS :25.000000 MICROF *
STANDARD VALUE OF CAPACITIVE REACTANCE IS :127.272728 OHMS

MACHINE LOSSES

IRON FLUX IN THE STATOR TEETH IS:2.278572 WEBERS
IRON LOSS IN THE STATOR :22.109751 WATTS
IRON LOSS IN THE ROTOR :22.109751 WATTS
COPPER LOSS IN THE STATOR:75.742546 WATTS
COPPER LOSS IN THE ROTOR:1.173802 WATTS
COPPER LOSS IN THE END RING IS :0.433118 WATTS
IRON LOSS IN THE MACHINE:44.219501 WATTS
COPPER LOSS IN THE MACHINE:77.349464 WATTS
FRICTION AND WINDAGE LOSSES IN THE MACHINE:3.677500 WATTS
LOSSES IN THE MACHINE IS:125.246466 WATTS

MACHINE CHARACTERSTIC DETAILS

START OF START WDG WITH CAPACITOR IN SERIES IS:109.684175 AMPS
NO-LOAD ROTOR CURRENT IN MAIN WDG IS:26.717461 AMPS
NO-LOAD ROTOR CURRENT IN AUX1 WDG IS:2.096924 AMPS
NO-LOAD STARTING CURRENT IS:28.814384 AMPS
NO-LOAD STARTING TORQUE IS:0.129035 NEWTON-METER
NO-LOAD SHORT CKT CURRENT IS:27.615698 AMPS
NO-LOAD SHORT CKT POWER FACTOR IS:0.508187
NO-LOAD SPEED IS GIVEN BY:519.127136
CALCULATED OUTPUT OF THE MACHINE IS :393.880676
CALCULATED EFFICIENCY OF MACHINE IS:75.873642

NO LOAD CHARACTERSTICS

NO-LOAD BRUSHING CURRENT IS :7.140621 AMPS
NO-LOAD WINDING LOSS IS :47.897003 WATTS
NO-LOAD CURRENT IS :7.143657 AMPS
NO-LOAD POWER FACTOR IS :0.029164

MMF VALUES OF MACHINE

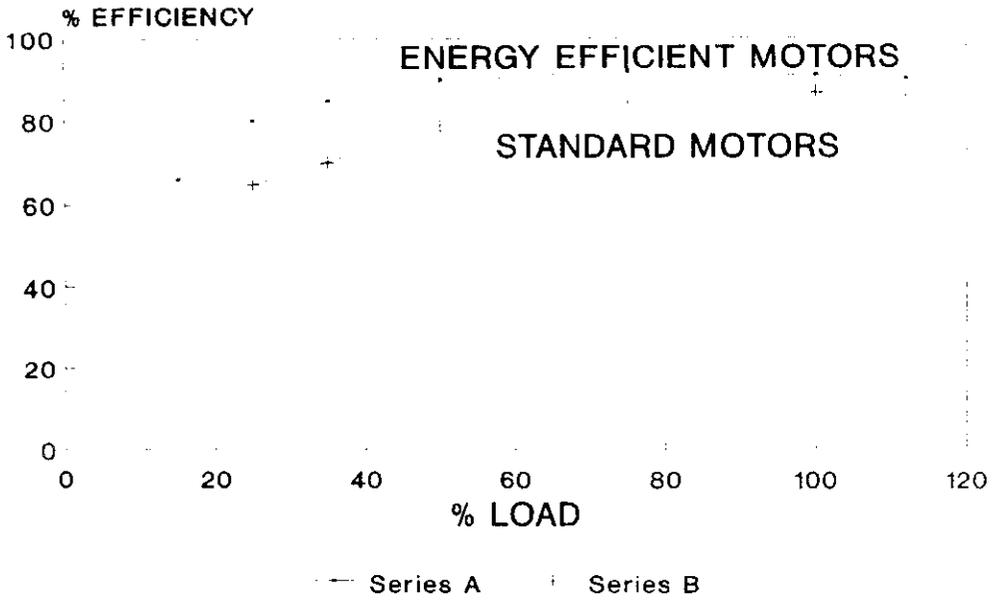
MMF IN THE AIR GAP IS:340.325664 AMPS
MMF IN THE STATOR TEETH IS:577.956848 AMPS
MMF IN THE ROTOR TEETH IS :11.703815 AMPS
MMF IN THE STATOR CORE IS :16.368261 AMPS
MMF IN THE ROTOR CORE IS :6.280798 AMPS
TOTAL MMF IN THE MACHINE IS:952.635376 AMPS

TEMPERATURE RISE

WINDING LOSS DISSIPATED BY STATOR SURFACE IS :67.762611 WATTS
WINDING LOSS DISSIPATION :2.481262 WATTS
TEMPERATURE RISE OF MACHINE IS :27.309740 DEGREE
TEMPERATURE RISE IS WITHIN THE LIMIT
*** COMPLETE ALL PARAMETER VALUES MOTOR*****

*** THANK YOU FOR USING SOFTWARE*****

COMP OF E.EFF MOTORS WITH STD MOTORS LOAD Vs EFFICIENCY CURVE





—●—
CONCLUSION
—●—

C H A P T E R - 8

CONCLUSION

In this project, an energy efficient 0.25 H.P,1440 rpm single phase squirrel cage induction motor has been designed, fabricated and tested successfully.

A computer program in C language has been developed for the design of the motor.

The motor is made energy efficient by reducing the losses. The losses were reduced by increasing the conductor area and core area and using better grade steel laminations. An increase of 10% in efficiency has been achieved.

These energy efficient motors can be widely used in domestic appliances like

- (1) Wet Grinders
- (2) Small pumps
- (3) Fans, etc.,

Appendix
